

# Digital Filters

## Music 171: Introduction to Delay and Filters

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- **Filter:** any medium through which a signal passes.
- Typically, a filter modifies the signal in some way:
  - audio speakers / headphones
  - rooms / acoustic spaces
  - musical instruments
- A *digital* filter is a formula for going from one digital signal (input  $x(n)$ ) to another (output  $y(n)$ ):

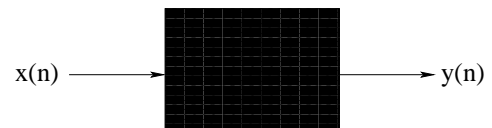
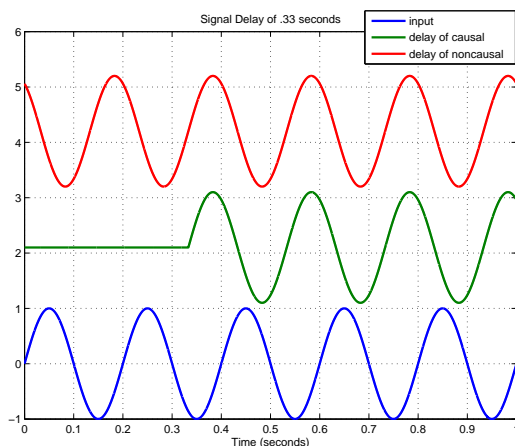


Figure 1: A black box filter.

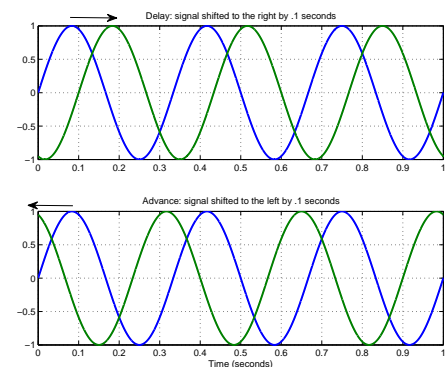
## Inside the Black Box—Pure Delay

- Digital filters typically involve signal *delay*.
- Delaying an audio signal is to
  - move it (earlier/later) in time;
  - change the **phase** of signal (the value at time=0).



## Time shifting a signal

- When a signal can be expressed in the form
 
$$y(n) = x(n - M),$$
 $y(n)$  is a *delayed* (time-shifted) version of  $x(n)$ .



- $y(n) = x(n - M)$ :  $x(n)$  is **delayed**  $M$  samples:
  - shift is to the **right** on the time axis.
- $y(n) = x(n + M)$ :  $x(n)$  is **advanced**  $M$  samples:
  - shift is to the **left** on the time axis.

## The Delay Line

- The delay line is a functional unit that models *acoustic propagation delay*.
- It is a fundamental building block of *delay effects processors*.
- The function of a delay line is introduce a time delay of  $M$  samples or

$$\tau = M/f_s \text{ seconds}$$

between its input and output.

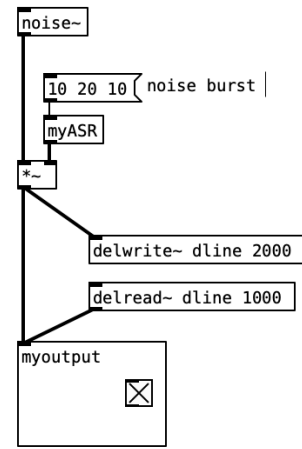


$$y(n) = x(n - M), \quad n = 0, 1, 2, \dots$$

## Time shift and addition

- Other than possible silence, there is no audible effect of a *pure* delay.

$x(n)$  and  $x(n - M)$  sound the same



- Change arises, however, when a signal  $x(n)$  is added to a delayed version of itself  $x(n - M)$ :

$$y(n) = x(n) + x(n - M)$$

## A Running Averager

- Consider a simple case where  $M = 1$ .

$$y(n) = x(n) + x(n - 1).$$

- (Dividing by 2), this filter **averages** adjacent samples.
  - that is, output  $y(n)$  is a *running average* of input  $x(n)$  **with a gain of 2**.

This filter takes the average of two adjacent samples.

## Intuitive Analysis at Low Frequencies

- Consider input at 0 Hz (lowest possible frequency):

$$x_1(n) = [A, A, A, \dots].$$

(at 0 Hz there is no change from sample to sample).

- The output is

$$\begin{aligned} y(n) &= x_1(n) + x_1(n - 1) \\ &= [A, A, A, \dots] \\ &\quad + [0, A, A, A, \dots] \\ &= [A, 2A, 2A, 2A, \dots] \\ &\approx 2x_1(n) \quad (\text{except 1st sample}). \end{aligned}$$

The filter has a gain of 2 at the lowest frequency.

## Intuitive Analysis at High Frequencies

- Consider input at  $\frac{f_s}{2}$  Hz (highest possible frequency):

$$x_2(n) = [A, -A, A, -A, \dots].$$

(maximum change from sample to sample).

- The output of the filter is

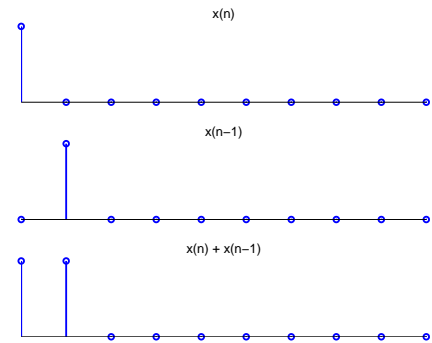
$$\begin{aligned} y(n) &= x_2(n) + x_2(n-1) \\ &= [A, -A, A, -A, \dots] \\ &\quad + [0, A, -A, A, \dots] \\ &= [A, 0, 0, 0, \dots] \\ &\approx 0x_2(n) \quad (\text{except 1st sample}). \end{aligned}$$

The filter has a gain of 0 the highest frequency.

- A filter that boosts low frequencies while attenuating higher frequencies is called a **lowpass filter**.

## What about frequencies in between?

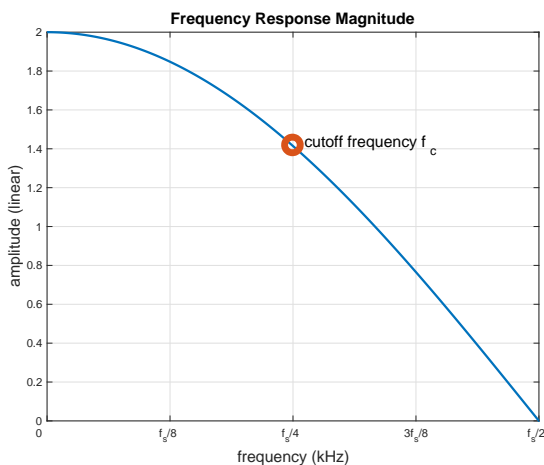
- Filter behaviour can be determined
  - using sinusoids at every possible frequency between 0 and  $f_s/2$  Hz;
  - using an input signal **that contains all frequency components** and check just once!
- Impulse**: signal with the broadest possible spectrum.



- Impulse Response (IR)**: response to an impulse (e.g. irCave.wav).

## Simple Lowpass Frequency Response

- Frequency response**:
  - spectrum of the impulse response;
  - shows how filter modifies frequency components.
- Frequency response of  $y(n) = x(n) + x(n-1)$ :



## Changing Filter Coefficients

- The **difference** (instead of the sum) of adjacent samples:

$$y(n) = x(n) - x(n-1).$$

is like changing the *coefficient* of  $x(n-1)$  to -1.

- At 0 Hz:

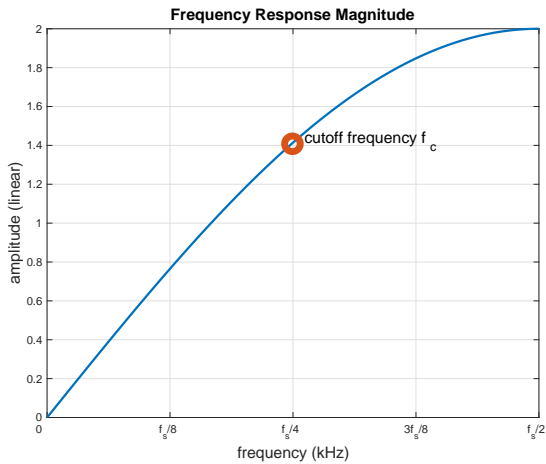
$$\begin{aligned} y(n) &= x_1(n) - x_1(n-1) \\ &= [A, A, A, \dots] \\ &\quad - [0, A, A, A, \dots] \\ &= [A, 0, 0, 0, \dots] \\ &\approx 0x_1(n). \end{aligned}$$

- At  $f_s/2$  Hz:

$$\begin{aligned} y(n) &= x_2(n) - x_2(n-1) \\ &= [A, -A, A, -A, \dots] \\ &\quad - [0, A, -A, A, \dots] \\ &= [A, -2A, 2A, -2A, \dots] \\ &\approx 2x_2(n). \end{aligned}$$

## Simple Highpass Frequency Response

- Frequency response shows a highpass filter.



- Notice the same cutoff frequency as simple lowpass.

## Notch Filter

- Changing the delay of the second term (and adding):

$$y(n) = x(n) + x(n - 2),$$

has output

- at 0 Hz ( $x_1(n) = [A, A, A, \dots]$ ):

$$\begin{aligned} y(n) &= [A, A, A, \dots] + [0, 0, A, A, \dots] \\ &= [A, A, 2A, 2A, \dots] \approx 2x_1(n). \end{aligned}$$

- at  $f_s/2$  Hz ( $x_2(n) = [A, -A, A, -A, \dots]$ ):

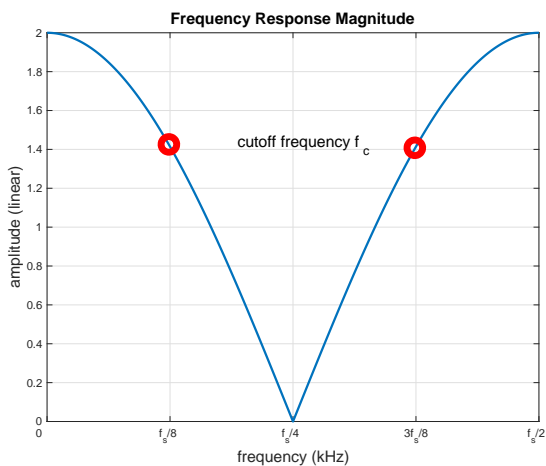
$$\begin{aligned} y(n) &= [A, -A, A, -A, \dots] + [0, 0, A, -A, A, \dots] \\ &= [A, -A, 2A, -2A, \dots] \approx 2x_2(n) \end{aligned}$$

- This filter boosts both low and high frequencies!
- Output at  $f_s/4$  Hz ( $x_3(n) = [A, 0, -A, 0, A, 0, \dots]$ ):

$$\begin{aligned} y(n) &= x_3(n) + x_3(n - 2) \\ &= [A, 0, -A, 0, A, \dots] + [0, 0, A, 0, -A, 0, \dots] \\ &= [A, 0, 0, 0, \dots] \approx 0x_3(n) \end{aligned}$$

## Simple Notch Frequency Response

- Frequency response shows a *notch* filter.



- Notice cutoff frequency is half of that for lowpass.

## Bandpass Filter

- Changing coefficient of  $x(n - 2)$  to -1:

$$y(n) = x(n) - x(n - 2),$$

yields output

- at 0 Hz:

$$\begin{aligned} y(n) &= [A, A, A, \dots] - [0, 0, A, A, \dots] \\ &= [A, A, 0, 0, \dots] \approx 0x(n). \end{aligned}$$

- at  $f_s/2$  Hz:

$$\begin{aligned} y(n) &= [A, -A, A, -A, \dots] - [0, 0, A, -A, A, \dots] \\ &= [A, -A, 0, 0, \dots] \approx 0x(n). \end{aligned}$$

- at  $f_s/4$  Hz:

$$\begin{aligned} y(n) &= [A, 0, -A, 0, A, \dots] - [0, 0, A, 0, -A, 0, \dots] \\ &= [A, 0, -2A, 0, 2A, 0, \dots] \approx 2x(n). \end{aligned}$$

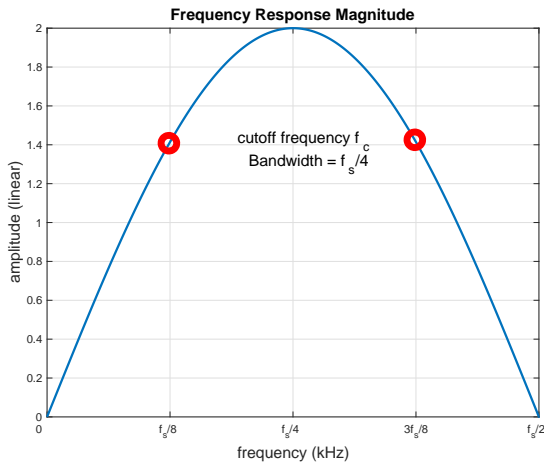
- Attenuation is at 0 and  $f_s/2$  Hz, and boosts  $f_s/4$

## Bandpass Filter Frequency Response

- The filter frequency (amplitude) response for

$$y(n) = x(n) - x(n - 2]$$

shows it is a **bandpass filter**.



- The bandwidth is determined by the frequency separation between the two cutoff points.

## Plots of simple filters

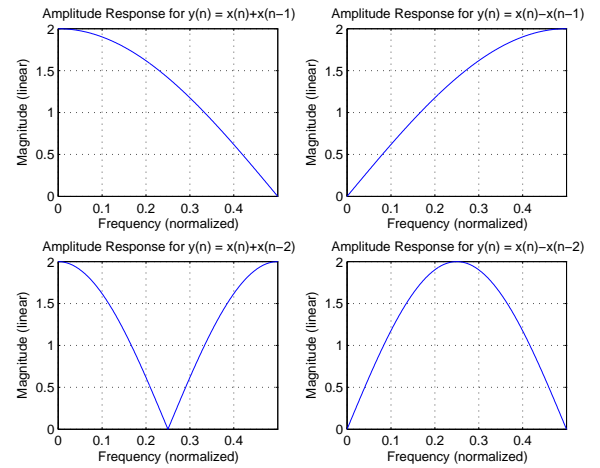


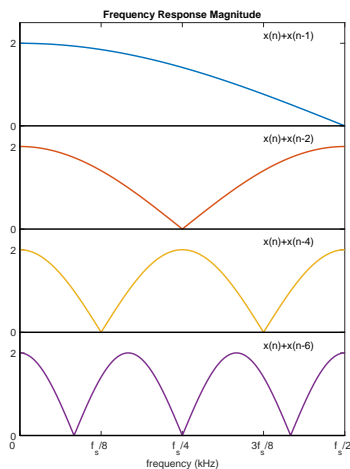
Figure 2: Amplitude Responses for simple filters

## Increasing the phase delay

- Make the delay of the  $2^{nd}$  term variable:

$$y(n) = x(n) + x(n - M]$$

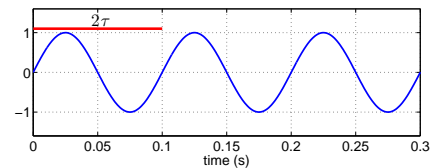
- Effects of increasing the  $M$  (the filter order):



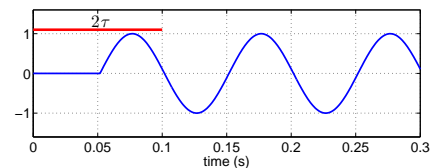
- Notice regularly spaced peaks and notches.
- Notches at odd harmonics of **what frequency?**

## Cancellation at Notch Frequencies

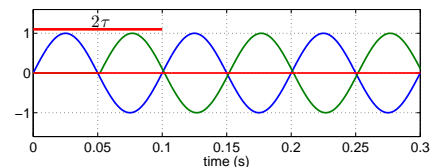
- Consider a sinusoid at  $f = 1/(2\tau)$  (period of  $2\tau$ ):



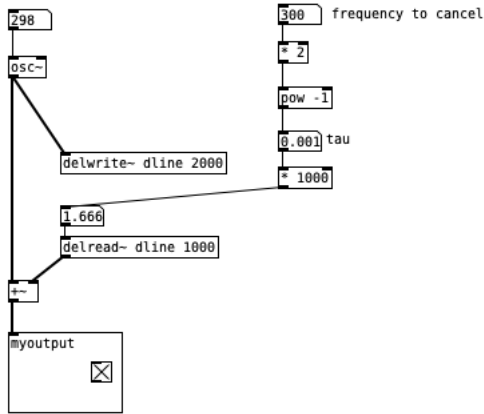
- Delaying that sinusoid by  $\tau$  (1/2 a period) yields:



- Summing with original yields complete cancellation:



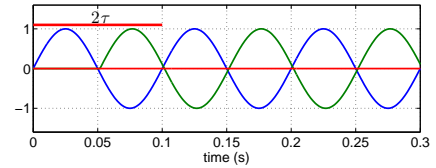
## Listen to Cancellation in Pd



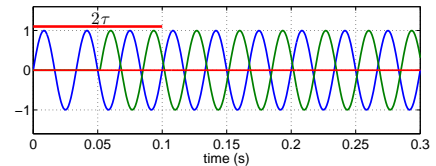
- comb11.21.19.pd

## Cancellation at Odd Harmonics

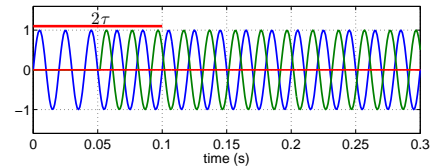
- Adding to a sinusoid at  $f = 1/(2\tau)$  a version of itself delayed by  $\tau$  yields cancellation at  $f = 1/(2\tau)$ ,



- but also at  $f = 3/(2\tau)$ ,



- and at  $5/(2\tau)$



- and at all **odd** harmonics of  $f = 1/(2\tau)$ .

## Relating $\tau$ to delay of $M$ samples

- For  $y(n) = x(n) + x(n - M)$  delay is  $M$  samples or

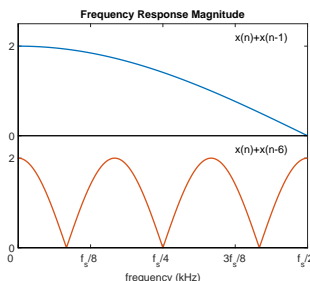
$$\tau = \frac{M}{f_s} \text{ seconds.}$$

- There is complete attenuation (notch) at frequency

$$f = \frac{1}{2\tau} = \frac{1}{2M/f_s} = \frac{f_s}{2M}$$

and at odd harmonics  $3f, 5f, \dots$  (up to Nyquist limit).

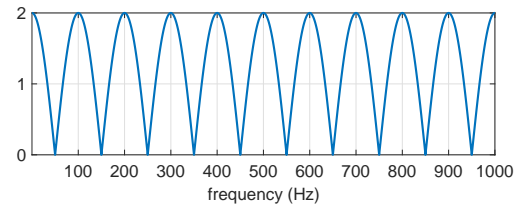
- For  $M = 1$  (lowpass) there is 1 notch at  $f_s/2$ .



- For  $M = 6$  there are notches at  $f_s/12, f_s/4, 5f_s/12$ .

## Feedforward Comb Filter

- Regular (comb-like) spacing of peaks/notches suggests **harmonics** of a fundamental frequency  $f_0$ .



- If notches are at **odd** harmonics of

$$f_n = \frac{1}{2\tau},$$

then peaks are at harmonics of

$$f_0 = 2f_n = \frac{2}{2\tau} = \frac{1}{\tau}.$$

- For a desired fundamental (sounding) frequency  $f_0$ :

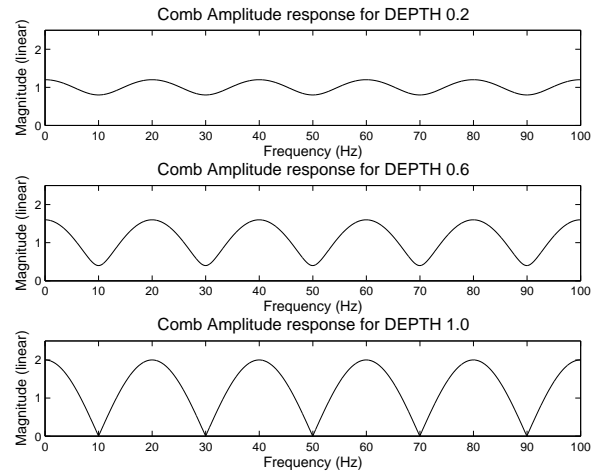
$$\tau = \frac{1}{f_0} \text{ seconds} \quad \text{OR} \quad M = \frac{f_s}{f_0} \text{ samples.}$$

## Feedforward Comb Filter in Pd

- see `ffcomb2.pd`.

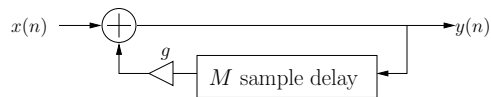
## Feedforward Comb Coefficient

- Introducing a coefficient allows for control of cancellation amount and the **depth** of notches:



## The Feedback Comb Filter

- What happens when the output of a delay line is multiplied by gain  $g < 1$  then fed back to the input?



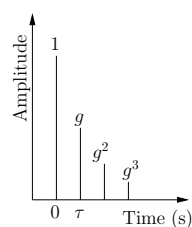
- The difference equation for this filter is

$$y(n) = x(n) + gy(n - M),$$

- If the input to the filter is an impulse

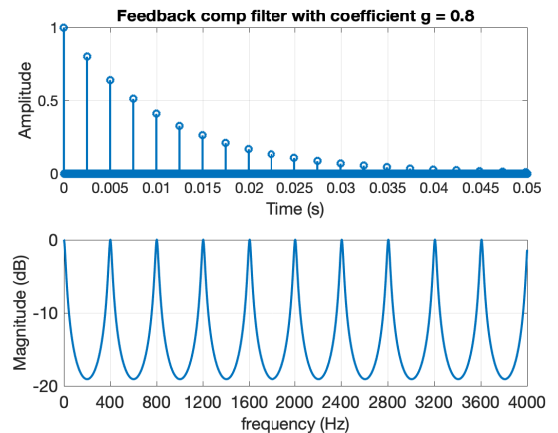
$$x(n) = \{1, 0, 0, \dots\}$$

the output (impulse response) will be ...



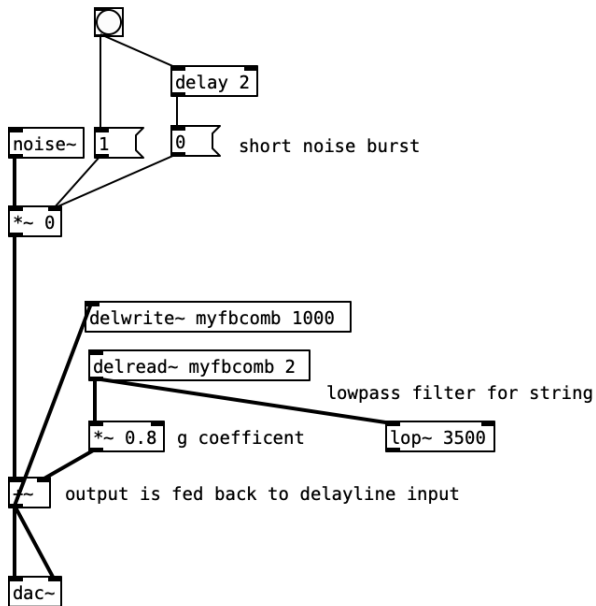
## Feedback Comb Filter Frequency

- Pulses are equally spaced in time at interval  $\tau = M/f_s$  seconds.
- It is periodic and will sound at frequency  $f_0 = 1/\tau$ .



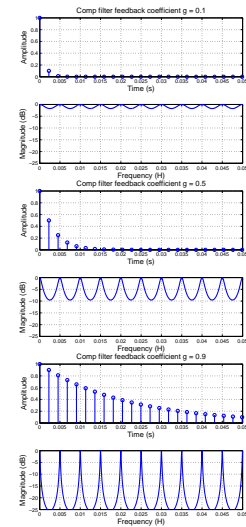
- Spacing between the maxima of the comb “teeth” is equal to the natural frequency  $f_0$ .

## Feedback Comb Filter in Pd



## Effect of the Feedback coefficient

- Minima depth and maxima height controlled by  $g$ .
- Values closer to 1 yield more extreme max/min.



## Comb Filter Decay Rate

- The response decays exponentially as determined by the loop time and gain factor  $g$ .
- Values of  $g$  nearest 1 yield the longest decay times.
- To obtain a desired decay time,  $g$  may be approximated by

$$g = 0.001^{\tau/T_{60}}$$

where

$\tau$  = the loop time

$T_{60}$  = the time to decay by 60dB

and 0.001 is the level of the signal at 60dB down.

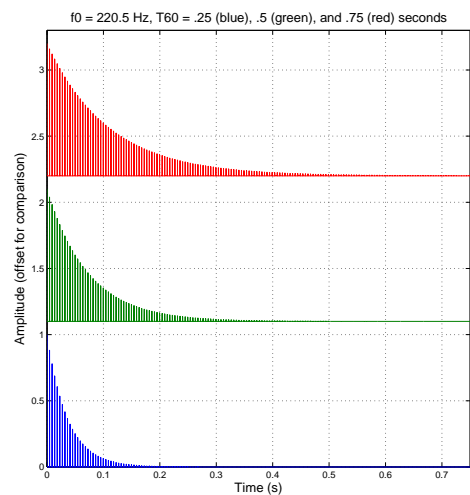


Figure 3: Comb filter impulse responses with a changing the decay rate.



## A very simple string model

- A very simple string model can be implemented using a single delay line and our simple first-order low pass filter to model frequency-dependent loss.

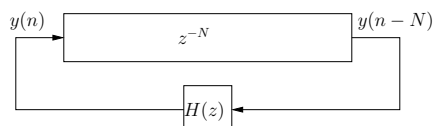


Figure 4: A very simple model of a rigidly terminated string.

- All losses have been *lumped* to a single observation point in the delay line, and approximated with our first-order simple low-pass filter

$$y(n) = x(n) + x(n - 1)$$

- Different sounds can be created by changing this filter.
- The Karplus-Strong Algorithm may be interpreted as a **feedback comb filter** (with lowpassed feedback) or a simplified **digital waveguide** model.
- How do you *pluck* the string? (noise burst.)